# **Digital Communications Laboratory**

**Dr. Mohammad Javad Emadi**

**Pouya Agheli**

# **Chapter Two**

**Analog Modulations and Demodulations**



**Electrical Engineering Department Amirkabir University of Technology Autumn 2020**

**E-mail: https://bit.ly/dclabAUT digitalcomslab@gma il.com digitalcomslab@gmail.com**

**Section A**

**Amplitude Modulations and Demodulations**

## AM-DSB-SC Modulation Slide 3 of 30

#### Expressions:

$$
s_i(t) = A_i \cos(2\pi f_i t) = A_i \cos(\omega_i t)
$$
  

$$
s_c(t) = A_c \cos(2\pi f_c t) = A_c \cos(\omega_c t)
$$
  

$$
s_{am-dsb-sc}(t) = A_i \cos(\omega_i t) A_c \cos(\omega_c t)
$$
  

$$
s_{am-dsb-sc}(t) = \frac{A_i A_c}{2} \Big( \cos(\omega_c - \omega_i)t + \cos(\omega_c + \omega_i)t \Big)
$$

Modulator:





## **AM-DSB-TC Modulation**

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#### Modulator:



Expressions:

$$
\begin{cases}\ns_i(t) = A_i \cos(2\pi f_i t) = A_i \cos(\omega_i t) \\
s_c(t) = A_c \cos(2\pi f_c t) = A_c \cos(\omega_c t)\n\end{cases}
$$

$$
s_{am-dsb-tc}(t) = \left[A_o + A_i \cos(\omega_i t)\right] A_c \cos(\omega_c t)
$$

$$
s_{am-dsb-tc}(t) = A_o \left[ 1 + m \cos(\omega_i t) \right] A_c \cos(\omega_c t)
$$

$$
= A_o A_c \cos(\omega_c t) + \frac{A_o A_c m}{2} \left( \cos(\omega_c - \omega_i)t + \cos(\omega_c + \omega_i)t \right)
$$



## **AM-SSB Modulation**

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#### Expressions:

$$
s_{am-ssb}(t) = s_i(t) \mathfrak{Re}\bigg[s_c(t)\bigg] \mp \overline{s_i(t)} \mathfrak{Im}\bigg[s_c(t)\bigg]
$$

 $s_i(t) = A_i \cos(2\pi f_i t) = A_i \cos(\omega_i t)$  $s_c(t) = A_c \cos(2\pi f_c t) + A_c \sin(2\pi f_c t) = A_c \cos(\omega_c t) + A_c \sin(\omega_c t)$ 

 $s_{am-ssb}(t) = A_i \cos(\omega_i t) A_c \cos(\omega_c t) \mp A_i \sin(\omega_i t) A_c \sin(\omega_c t)$ 

$$
s_{am-ssb}(t) = \frac{A_i A_c}{2} \left( \cos(\omega_c - \omega_i)t + \cos(\omega_c + \omega_i)t \right)
$$

$$
= \frac{A_i A_c}{2} \left( \cos(\omega_c - \omega_i)t - \cos(\omega_c + \omega_i)t \right)
$$





## **AM-VSB Modulation**



Expression:

$$
s_{am-vsb}(t) = BPF\left\{ \left[ A_o + s_i(t) \right] \times s_c(t) \right\}
$$

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#### Evaluation Table:

![](_page_6_Picture_77.jpeg)

**Note:** Each technique has some advantages and disadvantages, so based on deployment factor, a modulation should be selected.

**Section B**

**Angle Modulations and Demodulations**

## **Conventional PM and FM**

#### Expressions:  $\leftarrow 2 \times 10^{-4}$  $m(t)$  $\mathbf{\hat{m}}(t)$ 20,000  $\vert$  1  $t \rightarrow$  $\varphi_{\text{PM}}(t) = A \cos[\omega_c t + k_p m(t)]$  $t \rightarrow$ PM:  $-20,000$ Instantaneous angular  $\varphi_{\rm FM}(t)$  $\varphi_{\rm PM}(t)$ frequency for PM:  $t \rightarrow$ Instantaneous angular  $\omega_i(t) = \omega_c + k_f m(t)$ frequency for FM:

$$
\text{FM: } \left| \varphi_{\text{FM}}(t) = A \cos \left[ \omega_c t + k_f \int_{-\infty}^t m(\alpha) d\alpha \right] \right|
$$

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## **Conventional FM**

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**Time** 

Time

**Time** 

 $s_i(t)$ 

 $c(t)$ 

## Expressions:

$$
s_{f_m}(t) = A_c \cos\left(\omega_c t + \frac{\theta_{f_m}(t)}{\theta_{f_m}(t)}\right) = A_c \cos\left(\omega_c t + 2\pi K_{f_m} \times \int_{-\infty}^t s_i(t) dt\right)
$$

$$
s_i(t) = A_i \cos(2\pi f_i t) = A_i \cos(\omega_i t)
$$
  

$$
s_{fm}(t) = A_c \cos(\omega_c t + \beta_{fm} \sin(\omega_i t))
$$

$$
\omega_i = 2\pi f_i \qquad \omega_c = 2\pi f_c
$$

Modulator:

![](_page_9_Figure_7.jpeg)

INFORMATION<br>SIGNAL

Magnitude

Magnitude<br>
A

п.

Information signal is a baseband signal;

a sum of sinusoids

When there is no control input, the signal output<br>by the VCO is a pure cosinusoid

Implitude

mplitude

Frequency (kHz)

**Analog Phased-Lock Loop (APLL) Section C**

## **Analog PLL**

Free-running angular frequency VCO: Instantaneous angular  $\omega(t) = \omega_c + ce_o(t)$  $\phi_o(t) = ce_o(t)$  (A) frequency: Output signal:  $B \cos[\omega_c t + \theta_o(t)]$ 

![](_page_11_Figure_2.jpeg)

#### Loop Filter:

Input multiplied signal:

$$
AB \sin(\omega_c t + \theta_i) \cos(\omega_c t + \theta_o) = \frac{AB}{2} [\sin(\theta_i - \theta_o) + \sin(2\omega_c t + \theta_i + \theta_o)]
$$
  
Output signal:  $e_o(t) = h(t) * \frac{1}{2} AB \sin[\theta_i(t) - \theta_o(t)]$   

$$
= \frac{1}{2} AB \int_0^t h(t - x) \sin[\theta_i(x) - \theta_o(x)] dx
$$
 (B)

![](_page_11_Figure_6.jpeg)

$$
\xrightarrow{\text{(A)}} \dot{\theta}_o(t) = AK \int_0^t h(t-x) \sin \theta_e(x) dx
$$

Phase error:  $\theta_e(t) = \theta_i(t) - \theta_o(t)$ 

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**GNU Radio and SDR Section D**

#### Preferred GRC Blocks:

![](_page_13_Picture_92.jpeg)

**Note:** You may need other essential blocks.

#### Transmitter's Flow -Graph:

![](_page_14_Figure_2.jpeg)

Channel's Flow-Graph:

![](_page_15_Figure_2.jpeg)

### Receiver's Flow-Graph:

![](_page_16_Figure_2.jpeg)

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## **GRC Example for AM -DSB -TC**

### Signals' Figures in Time Series:

![](_page_17_Figure_2.jpeg)

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## **GRC Example for AM -DSB -TC**

#### Signals' Figures in Frequency Series:

![](_page_18_Figure_2.jpeg)

## **GRC Example for AM -DSB -TC**

## Signals' Waterfall Figures:

![](_page_19_Figure_3.jpeg)

## Preferred GRC Blocks:

![](_page_20_Picture_93.jpeg)

**Note:** You may need other essential blocks.

#### Transmitter's Flow-Graph:

![](_page_21_Figure_2.jpeg)

## Channel's Flow-Graph:

![](_page_22_Figure_2.jpeg)

Receiver's Flow-Graph:

![](_page_23_Figure_2.jpeg)

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### Signals' Figures in Time Series:

![](_page_24_Figure_2.jpeg)

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#### Signals' Figures in Frequency Series:

![](_page_25_Figure_2.jpeg)

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## Signals' Waterfall Figures:

![](_page_26_Figure_3.jpeg)

**nooelec:** We designed this SDR from the ground up in order to develop the best low-cost SDR in existence.

**Frequency Range:** 25MHz - 1750MHz **Phase noise @1kHz offset:** -138dBc/Hz (or better) **Phase noise @10kHz:** -150dBc/Hz (or better) **Phase noise @100kHz:** -152dBc/Hz (or better)

![](_page_27_Picture_3.jpeg)

![](_page_27_Picture_4.jpeg)

# **Assignments**

**Session Two**

**Problem:**

Design AM-SSB via GNU Radio

**Due:** Oct. 13, 2020

# **Assignments**

**Session Three**

**Problem:**

Design Conventional PM via GNU Radio

**Due:** Oct. 20, 2020